

FEATURES

- Single Supply Operation
 - Input Voltage Range Extends to Ground
 - Output Swings to Ground while Sinking Current
- Pin Compatible to 1458 and 324 with Precision Specs
- *Guaranteed* Offset Voltage 150 μ V Max.
- *Guaranteed* Low Drift 2 μ V/ $^{\circ}$ C Max.
- *Guaranteed* Offset Current 0.8nA Max.
- *Guaranteed* High Gain
 - 5mA Load Current 1.5 Million Min.
 - 17mA Load Current 0.8 Million Min.
- *Guaranteed* Low Supply Current 500 μ A Max.
- Low Voltage Noise, 0.1Hz to 10Hz 0.55 μ Vp-p
- Low Current Noise—Better than OP-07, 0.07pA/ $\sqrt{\text{Hz}}$

APPLICATIONS

- Battery-Powered Precision Instrumentation
 - Strain Gauge Signal Conditioners
 - Thermocouple Amplifiers
 - Instrumentation Amplifiers
- 4mA–20mA Current Loop Transmitters
- Multiple Limit Threshold Detection
- Active Filters
- Multiple Gain Blocks

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DESCRIPTION

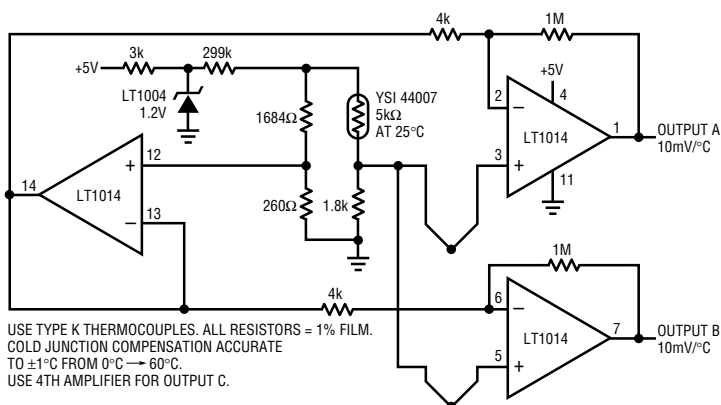
The LT[®]1014 is the first precision quad operational amplifier which directly upgrades designs in the industry standard 14-pin DIP LM324/LM348/OP-11/4156 pin configuration. It is no longer necessary to compromise specifications, while saving board space and cost, as compared to single operational amplifiers.

The LT1014's low offset voltage of 50 μ V, drift of 0.3 μ V/ $^{\circ}$ C, offset current of 0.15nA, gain of 8 million, common-mode rejection of 117dB and power supply rejection of 120dB qualify it as four truly precision operational amplifiers. Particularly important is the low offset voltage, since no offset null terminals are provided in the quad configuration. Although supply current is only 350 μ A per amplifier, a new output stage design sources and sinks in excess of 20mA of load current, while retaining high voltage gain.

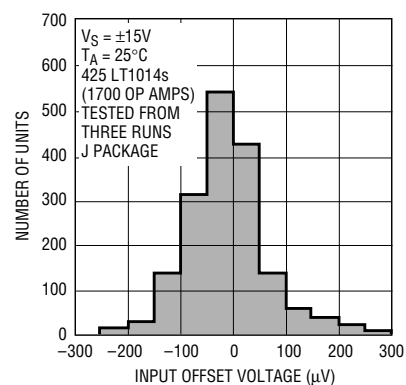
Similarly, the LT1013 is the first precision dual op amp in the 8-pin industry standard configuration, upgrading the performance of such popular devices as the MC1458/1558, LM158 and OP-221. The LT1013's specifications are similar to (even somewhat better than) the LT1014's.

Both the LT1013 and LT1014 can be operated off a single 5V power supply: input common-mode range includes ground; the output can also swing to within a few millivolts of ground. Crossover distortion, so apparent on previous single-supply designs, is eliminated. A full set of specifications is provided with ± 15 V and single 5V supplies.

3 Channel Thermocouple Thermometer



LT1014 Distribution of Offset Voltage



LT1013/LT1014

ABSOLUTE MAXIMUM RATINGS

Supply Voltage $\pm 22V$
 Differential Input Voltage $\pm 30V$
 Input Voltage Equal to Positive Supply Voltage
 5V Below Negative Supply Voltage
 Output Short-Circuit Duration Indefinite
 Storage Temperature Range
 All Grades $-65^{\circ}C$ to $150^{\circ}C$

Lead Temperature (Soldering, 10 sec.) $300^{\circ}C$
 Operating Temperature Range
 LT1013AM/LT1013M/
 LT1014AM/LT1014M $-55^{\circ}C$ to $125^{\circ}C$
 LT1013AC/LT1013C/LT1013D
 LT1014AC/LT1014C/LT1014D $0^{\circ}C$ to $70^{\circ}C$
 LT1013I/ LT1014I $-40^{\circ}C$ to $85^{\circ}C$

PACKAGE/ORDER INFORMATION

TOP VIEW OUTPUT A 1, -IN A 2, +IN A 3, V-(CASE) 4, +IN B 5, -IN B 6, OUTPUT B 7, V+ 8 H PACKAGE 8-LEAD TO-5 METAL CAN	ORDER PART NUMBER LT1013AMH LT1013MH LT1013ACH LT1013CH
TOP VIEW OUTPUT A 1, -IN A 2, +IN A 3, V- 4, +IN B 5, -IN B 6, OUTPUT B 7, V+ 8 J PACKAGE 8-LEAD CERAMIC DIP N PACKAGE 8-LEAD PLASTIC DIP	ORDER PART NUMBER LT1013AMJ8 LT1013MJ8 LT1013ACJ8 LT1013CJ8 LT1013ACN8 LT1013CN8 LT1013DN8 LT1013IN8
TOP VIEW OUTPUT A 1, -IN A 2, +IN A 3, V+ 4, +IN B 5, -IN B 6, OUTPUT B 7, V- 8, +IN C 9, -IN C 10, +IN D 11, -IN D 12, OUTPUT D 13, V+ 14 J PACKAGE 14-LEAD CERAMIC DIP N PACKAGE 14-LEAD PLASTIC DIP	ORDER PART NUMBER LT1014AMJ LT1014MJ LT1014ACJ LT1014CJ LT1014ACN LT1014CN LT1014DN LT1014IN

TOP VIEW +IN A 1, V- 2, +IN B 3, -IN B 4, OUTB 5, V+ 6, OUTA 7, -IN A 8 SO PACKAGE 8-LEAD PLASTIC SOIC NOTE: THIS PIN CONFIGURATION DIFFERS FROM THE STANDARD 8-PIN DUAL-IN-LINE CONFIGURATION	ORDER PART NUMBER LT1013DS8 LT1013IS8	PART MARKING 1013 1013I
TOP VIEW OUTPUT A 1, -IN A 2, +IN A 3, V+ 4, +IN B 5, -IN B 6, OUTPUT B 7, NC 8, NC 9, OUTPUT C 10, -IN C 11, +IN C 12, V- 13, +IN D 14, -IN D 15, OUTPUT D 16 SO PACKAGE 16-LEAD PLASTIC SOIC	ORDER PART NUMBER LT1014DS LT1014IS	PART MARKING LT1014DS LT1014IS

ELECTRICAL CHARACTERISTICS

$V_S = \pm 15V$, $V_{CM} = 0V$, $T_A = 25^{\circ}C$ unless otherwise noted

SYMBOL	PARAMETER	CONDITIONS	LT1013AM/AC LT1014AM/AC			LT1013C/D/I/M LT1014C/D/I/M			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{OS}	Input Offset Voltage	LT1013	—	40	150	—	60	300	μV
		LT1014	—	50	180	—	60	300	μV
		LT1013D/I, LT1014D/I	—	—	—	—	200	800	μV
	Long Term Input Offset Voltage Stability		—	0.4	—	—	0.5	—	$\mu V/Mo.$
I_{SO}	Input Offset Current		—	0.15	0.8	—	0.2	1.5	nA
I_B	Input Bias Current		—	12	20	—	15	30	nA
e_n	Input Noise Voltage	0.1Hz to 10Hz	—	0.55	—	—	0.55	—	$\mu Vp-p$
e_n	Input Noise Voltage Density	$f_0 = 10Hz$	—	24	—	—	24	—	$nV/\sqrt{Hz}$
		$f_0 = 1000Hz$	—	22	—	—	22	—	$nV/\sqrt{Hz}$
i_n	Input Noise Current Density	$f_0 = 10Hz$	—	0.07	—	—	0.07	—	$pA/\sqrt{Hz}$

ELECTRICAL CHARACTERISTICS

$V_S = \pm 15V$, $V_{CM} = 0V$, $T_A = 25^\circ C$ unless otherwise noted

SYMBOL	PARAMETER	CONDITIONS	LT1013AM/AC LT1014AM/AC			LT1013C/D/I/M LT1014C/D/I/M			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
	Input Resistance – Differential Common-Mode	(Note 1)	100 —	400 5	— —	70 —	300 4	— —	M Ω G Ω
A_{VOL}	Large Signal Voltage Gain	$V_O = \pm 10V$, $R_L = 2k$	1.5	8.0	—	1.2	7.0	—	V/ μV
		$V_O = \pm 10V$, $R_L = 600\Omega$	0.8	2.5	—	0.5	2.0	—	V/ μV
	Input Voltage Range		+13.5 –15.0	+13.8 –15.3	— —	+13.5 –15.0	+13.8 –15.3	— —	V V
CMRR	Common-Mode Rejection Ratio	$V_{CM} = +13.5V, -15.0V$	100	117	—	97	114	—	dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 2V$ to $\pm 18V$	103	120	—	100	117	—	dB
	Channel Separation	$V_O = \pm 10V$, $R_L = 2k$	123	140	—	120	137	—	dB
V_{OUT}	Output Voltage Swing	$R_L = 2k$	± 13	± 14	—	± 12.5	± 14	—	V
	Slew Rate		0.2	0.4	—	0.2	0.4	—	V/ μs
I_S	Supply Current	Per Amplifier	—	0.35	0.50	—	0.35	0.55	mA

Note 1: This parameter is guaranteed by design and is not tested. Typical parameters are defined as the 60% yield of parameter distributions of individual amplifiers; i.e., out of 100 LT1014s (or 100 LT1013s) typically 240 op amps (or 120) will be better than the indicated specification.

ELECTRICAL CHARACTERISTICS

$V_{S^+} = +5V$, $V_{S^-} = 0V$, $V_{OUT} = 1.4V$, $V_{CM} = 0V$, $T_A = 25^\circ C$ unless otherwise noted

SYMBOL	PARAMETER	CONDITIONS	LT1013AM/AC LT1014AM/AC			LT1013C/D/I/M LT1014C/D/I/M			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{OS}	Input Offset Voltage	LT1013	—	60	250	—	90	450	μV
		LT1014	—	70	280	—	90	450	μV
		LT1013D/I, LT1014D/I	—	—	—	—	250	950	μV
I_{OS}	Input Offset Current		—	0.2	1.3	—	0.3	2.0	nA
I_B	Input Bias Current		—	15	35	—	18	50	nA
A_{VOL}	Large Signal Voltage Gain	$V_O = 5mV$ to $4V$, $R_L = 500\Omega$	—	1.0	—	—	1.0	—	V/ μV
	Input Voltage Range		+3.5 0	+3.8 –0.3	— —	+3.5 0	+3.8 –0.3	— —	V V
V_{OUT}	Output Voltage Swing	Output Low, No Load	—	15	25	—	15	25	mV
		Output Low, 600 Ω to Ground	—	5	10	—	5	10	mV
		Output Low, $I_{SINK} = 1mA$	—	220	350	—	220	350	mV
		Output High, No Load	4.0	4.4	—	4.0	4.4	—	V
		Output High, 600 Ω to Ground	3.4	4.0	—	3.4	4.0	—	V
I_S	Supply Current	Per Amplifier	—	0.31	0.45	—	0.32	0.50	mA

ELECTRICAL CHARACTERISTICS $V_S = \pm 15V$, $V_{CM} = 0V$, $-55^\circ C \leq T_A \leq 125^\circ C$ unless otherwise noted

SYMBOL	PARAMETER	CONDITIONS		LT1013AM			LT1014AM			LT1013M/LT1014M			UNITS
				MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{OS}	Input Offset Voltage	$V_S = +5V$, $0V$; $V_O = +1.4V$ $-55^\circ C \leq T_A \leq 100^\circ C$ $V_{CM} = 0.1V$, $T_A = 125^\circ C$ $V_{CM} = 0V$, $T_A = 125^\circ C$	●	—	80	300	—	90	350	—	110	550	μV
			●	—	80	450	—	90	480	—	100	750	μV
			●	—	120	450	—	150	480	—	200	750	μV
			●	—	250	900	—	300	960	—	400	1500	μV
	Input Offset Voltage Drift	(Note 2)	●	—	0.4	2.0	—	0.4	2.0	—	0.5	2.5	$\mu V/^\circ C$
I_{OS}	Input Offset Current	$V_S = +5V$, $0V$; $V_O = +1.4V$	●	—	0.3	2.5	—	0.3	2.8	—	0.4	5.0	nA
			●	—	0.6	6.0	—	0.7	7.0	—	0.9	10.0	nA
I_B	Input Bias Current	$V_S = +5V$, $0V$; $V_O = +1.4V$	●	—	15	30	—	15	30	—	18	45	nA
			●	—	20	80	—	25	90	—	28	120	nA
A_{VOL}	Large Signal Voltage Gain	$V_O = \pm 10V$, $R_L = 2k$	●	0.5	2.0	—	0.4	2.0	—	0.25	2.0	—	V/ μV
CMRR	Common-Mode Rejection	$V_{CM} = +13.0V$, $-14.9V$	●	97	114	—	96	114	—	94	113	—	dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 2V$ to $\pm 18V$	●	100	117	—	100	117	—	97	116	—	dB
V_{OUT}	Output Voltage Swing	$R_L = 2k$ $V_S = +5V$, $0V$ $R_L = 600\Omega$ to Ground Output Low Output High	●	± 12	± 13.8	—	± 12	± 13.8	—	± 11.5	± 13.8	—	V
			●	—	6	15	—	6	15	—	6	18	mV
			●	3.2	3.8	—	3.2	3.8	—	3.1	3.8	—	V
			●	—	0.38	0.60	—	0.38	0.60	—	0.38	0.7	mA
I_S	Supply Current Per Amplifier	$V_S = +5V$, $0V$; $V_O = +1.4V$	●	—	0.34	0.55	—	0.34	0.55	—	0.34	0.65	mA
			●	—	0.34	0.55	—	0.34	0.55	—	0.34	0.65	mA

ELECTRICAL CHARACTERISTICS

$V_S = \pm 15V$, $V_{CM} = 0V$, $-40^\circ C \leq T_A \leq 85^\circ C$ for LT1013I, LT1014I, $0^\circ C \leq T_A \leq 70^\circ C$ for LT1013C, LT1013D, LT1014C, LT1014D unless otherwise noted

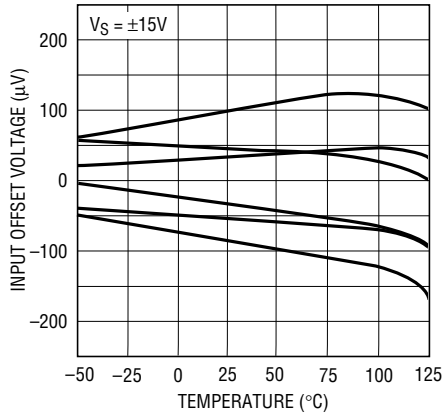
SYMBOL	PARAMETER	CONDITIONS		LT1013AC			LT1014AC			LT1013C/D/I LT1014C/D/I			UNITS
				MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V_{OS}	Input Offset Voltage	LT1013D/I, LT1014D/I $V_S = +5V$, $0V$; $V_O = 1.4V$ LT1013D/I, LT1014D/I $V_S = +5V$, $0V$; $V_O = 1.4V$	●	—	55	240	—	65	270	—	80	400	μV
			●	—	—	—	—	—	—	—	230	1000	μV
			●	—	75	350	—	85	380	—	110	570	μV
			●	—	—	—	—	—	—	—	280	1200	μV
	Average Input Offset Voltage Drift	(Note 2) LT1013D/I, LT1014D/I	●	—	0.3	2.0	—	0.3	2.0	—	0.4	2.5	$\mu V/^\circ C$
			●	—	—	—	—	—	—	—	0.7	5.0	$\mu V/^\circ C$
I_{OS}	Input Offset Current	$V_S = +5V$, $0V$; $V_O = 1.4V$	●	—	0.2	1.5	—	0.2	1.7	—	0.3	2.8	nA
			●	—	0.4	3.5	—	0.4	4.0	—	0.5	6.0	nA
I_B	Input Bias Current	$V_S = +5V$, $0V$; $V_O = 1.4V$	●	—	13	25	—	13	25	—	16	38	nA
			●	—	18	55	—	20	60	—	24	90	nA
A_{VOL}	Large Signal Voltage Gain	$V_O = \pm 10V$, $R_L = 2k$	●	1.0	5.0	—	1.0	5.0	—	0.7	4.0	—	V/ μV
CMRR	Common-Mode Rejection Ratio	$V_{CM} = +13.0V$, $-15.0V$	●	98	116	—	98	116	—	94	113	—	dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 2V$ to $\pm 18V$	●	101	119	—	101	119	—	97	116	—	dB
V_{OUT}	Output Voltage Swing	$R_L = 2k$ $V_S = +5V$, $0V$; $R_L = 600\Omega$ Output Low Output High	●	± 12.5	± 13.9	—	± 12.5	± 13.9	—	± 12.0	± 13.9	—	V
			●	—	6	13	—	6	13	—	6	13	mV
			●	3.3	3.9	—	3.3	3.9	—	3.2	3.9	—	V
			●	—	0.36	0.55	—	0.36	0.55	—	0.37	0.60	mA
I_S	Supply Current per Amplifier	$V_S = +5V$, $0V$; $V_O = 1.4V$	●	—	0.32	0.50	—	0.32	0.50	—	0.34	0.55	mA
			●	—	0.32	0.50	—	0.32	0.50	—	0.34	0.55	mA

Note 2: This parameter is not 100% tested.

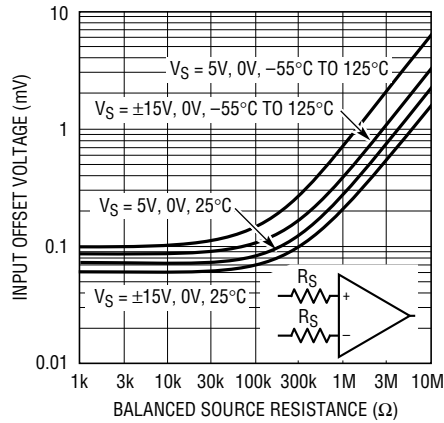
The ● denotes specifications which apply over the full operating temperature range.

TYPICAL PERFORMANCE CHARACTERISTICS

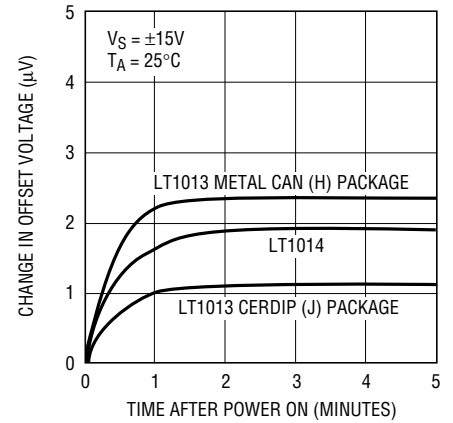
Offset Voltage Drift with Temperature of Representative Units



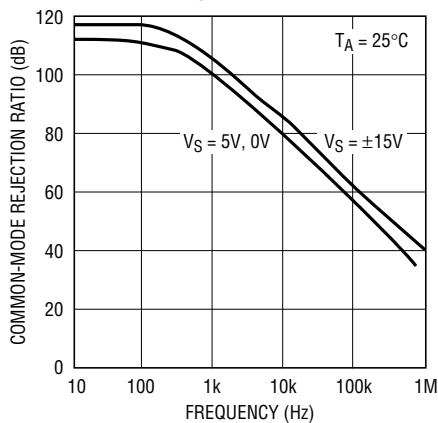
Offset Voltage vs Balanced Source Resistance



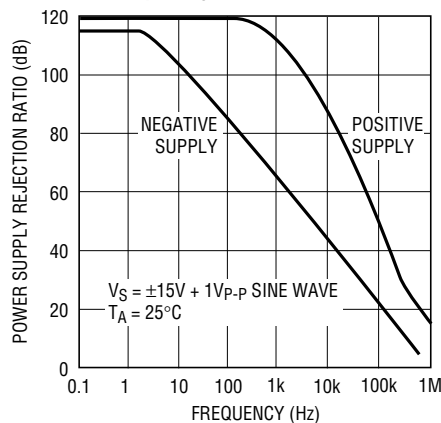
Warm-Up Drift



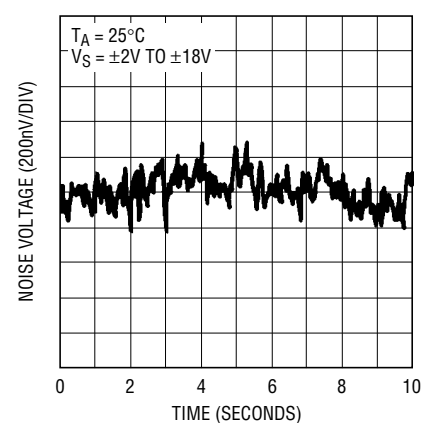
Common-Mode Rejection Ratio vs Frequency



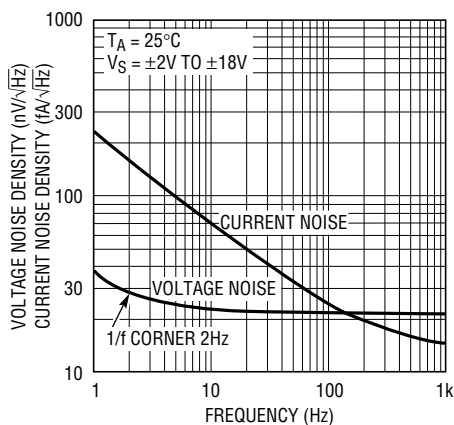
Power Supply Rejection Ratio vs Frequency



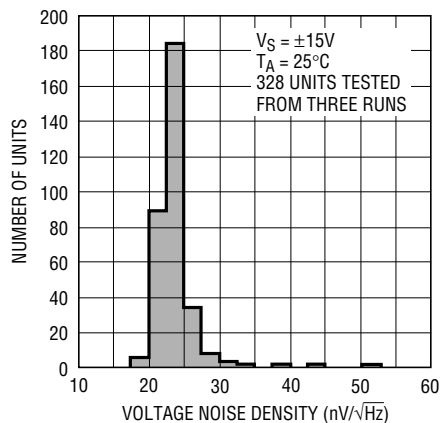
0.1Hz to 10Hz Noise



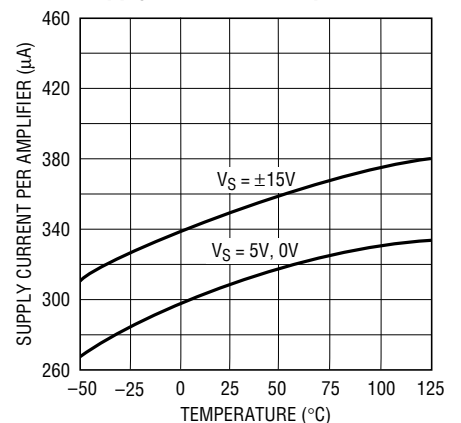
Noise Spectrum



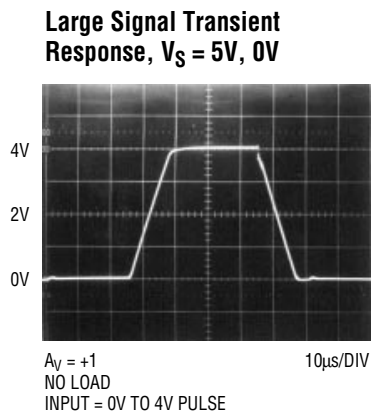
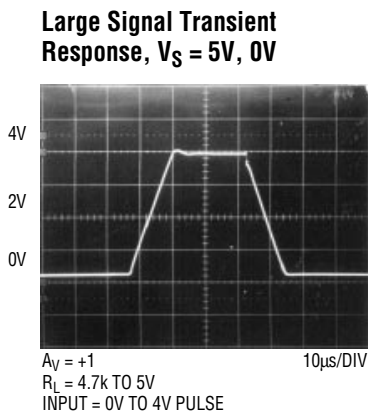
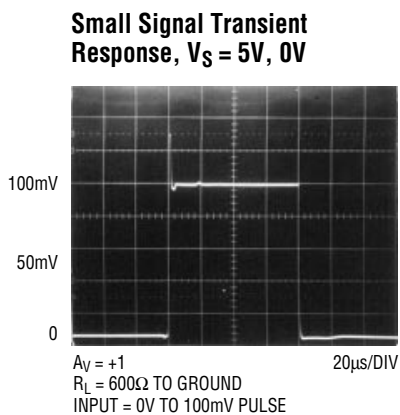
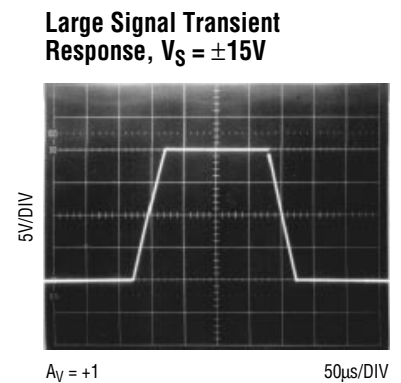
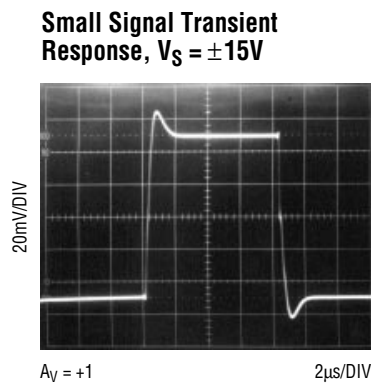
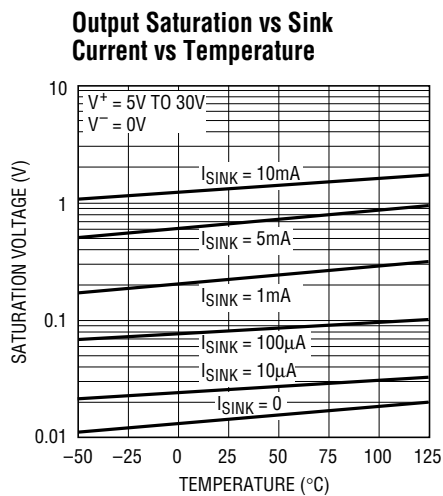
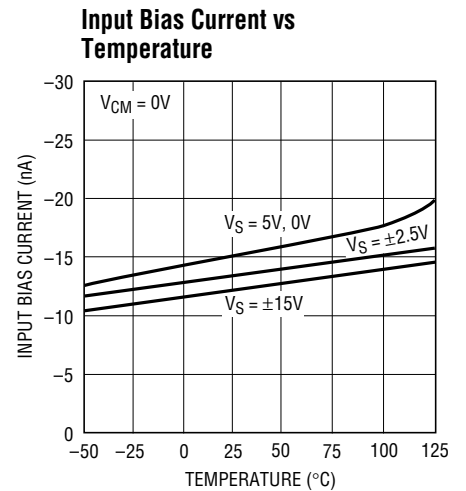
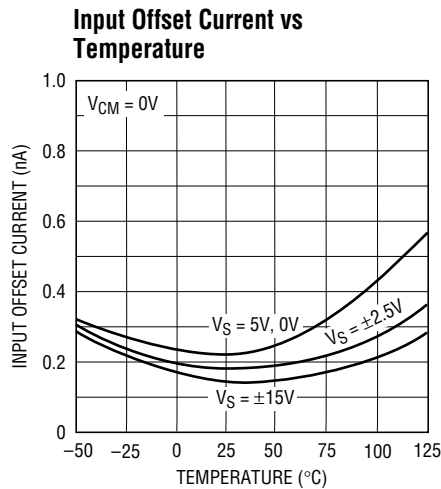
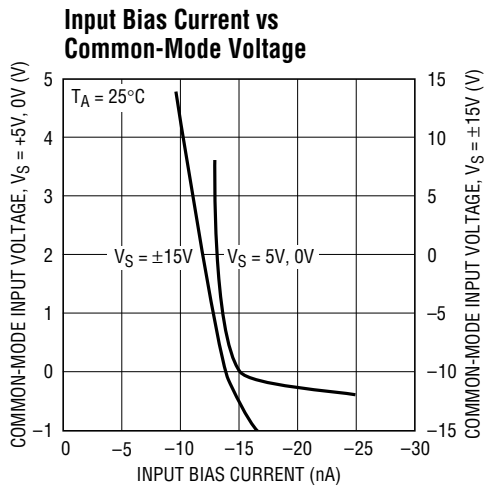
10Hz Voltage Noise Distribution



Supply Current vs Temperature

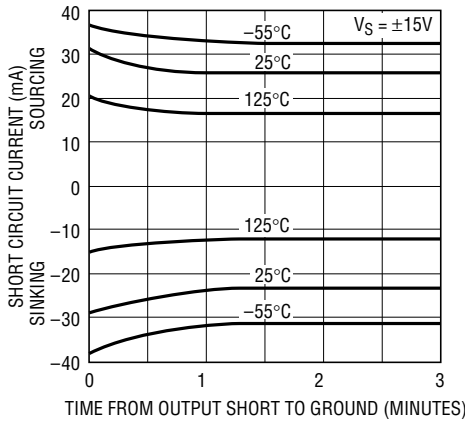


TYPICAL PERFORMANCE CHARACTERISTICS

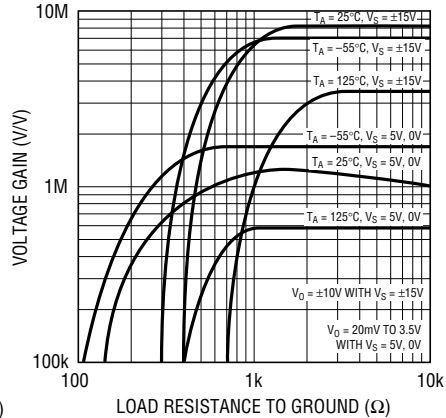


TYPICAL PERFORMANCE CHARACTERISTICS

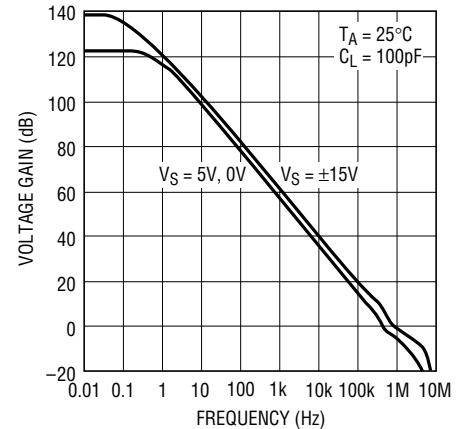
Output Short Circuit Current vs Time



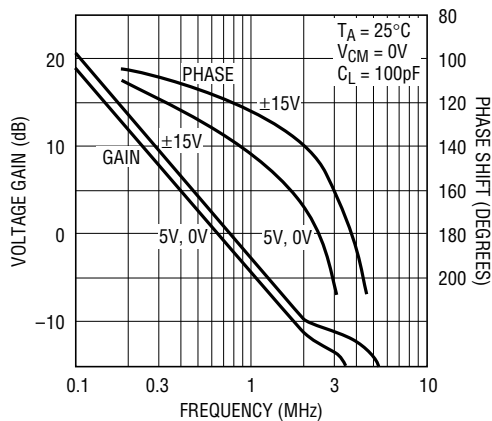
Voltage Gain vs Load Resistance



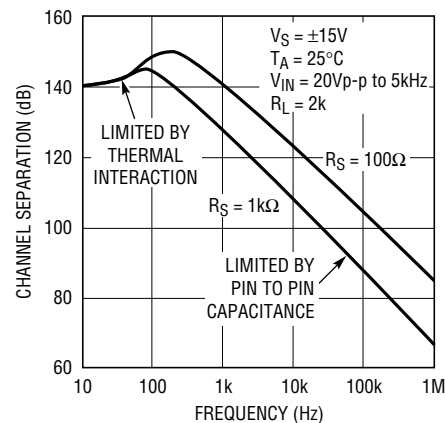
Voltage Gain vs Frequency



Gain, Phase vs Frequency



Channel Separation vs Frequency



APPLICATIONS INFORMATION

Single Supply Operation

The LT1013/1014 are fully specified for single supply operation, i.e., when the negative supply is 0V. Input common-mode range includes ground; the output swings within a few millivolts of ground. Single supply operation, however, can create special difficulties, both at the input and at the output. The LT1013/LT1014 have specific circuitry which addresses these problems.

At the input, the driving signal can fall below 0V—inadvertently or on a transient basis. If the input is more than

a few hundred millivolts below ground, two distinct problems can occur on previous single supply designs, such as the LM124, LM158, OP-20, OP-21, OP-220, OP-221, OP-420:

a) When the input is more than a diode drop below ground, unlimited current will flow from the substrate (V^- terminal) to the input. This can destroy the unit. On the LT1013/1014, the 400Ω resistors, in series with the input (see schematic diagram), protect the devices even when the input is 5V below ground.

APPLICATIONS INFORMATION

(b) When the input is more than 400mV below ground (at 25°C), the input stage saturates (transistors Q3 and Q4) and phase reversal occurs at the output. This can cause lock-up in servo systems. Due to a unique phase reversal protection circuitry (Q21, Q22, Q27, Q28), the LT1013/1014's outputs do not reverse, as illustrated below, even when the inputs are at -1.5V.

There is one circumstance, however, under which the phase reversal protection circuitry does not function: when the other op amp on the LT1013, or one specific amplifier of the other three on the LT1014, is driven hard into negative saturation at the output.

Phase reversal protection does not work on amplifier:

A when D's output is in negative saturation. B's and C's outputs have no effect.

B when C's output is in negative saturation. A's and D's outputs have no effect.

C when B's output is in negative saturation. A's and D's outputs have no effect.

D when A's output is in negative saturation. B's and C's outputs have no effect.

At the output, the aforementioned single supply designs either cannot swing to within 600mV of ground (OP-20) or cannot sink more than a few microamperes while swinging to ground (LM124, LM158). The LT1013/1014's all-NPN output stage maintains its low output resistance and high gain characteristics until the output is saturated.

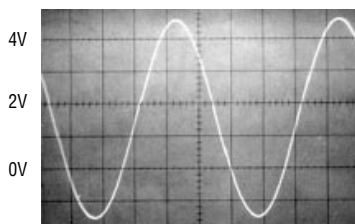
In dual supply operations, the output stage is crossover distortion-free.

Comparator Applications

The single supply operation of the LT1013/1014 lends itself to its use as a precision comparator with TTL compatible output:

In systems using both op amps and comparators, the LT1013/1014 can perform multiple duties; for example, on the LT1014, two of the devices can be used as op amps and the other two as comparators.

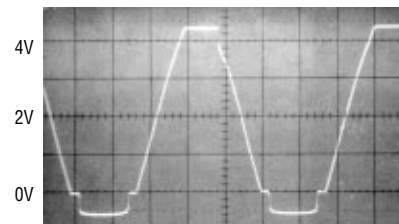
Voltage Follower with Input Exceeding the Negative Common-Mode Range



6Vp-p INPUT, -1.5V TO 4.5V

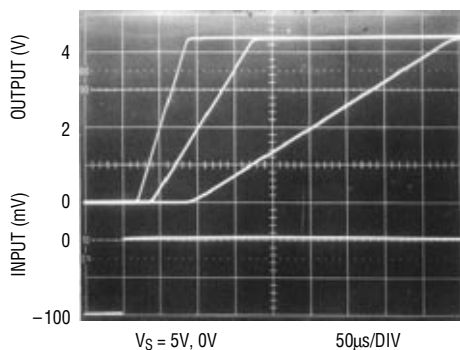


LM324, LM358, OP-20
EXHIBIT OUTPUT PHASE
REVERSAL

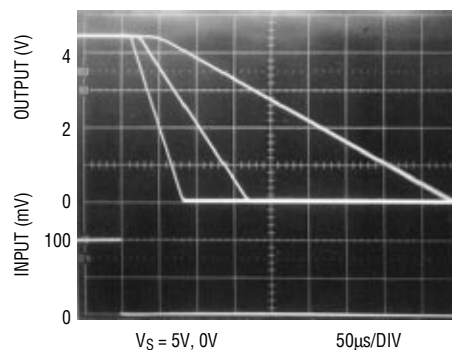


LT1013/LT1014
NO PHASE REVERSAL

Comparator Rise Response Time 10mV, 5mV, 2mV Overdrives



Comparator Fall Response Time to 10mV, 5mV, 2mV Overdrives



APPLICATIONS INFORMATION

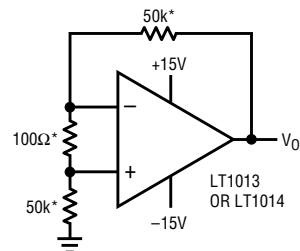
Low Supply Operation

The minimum supply voltage for proper operation of the LT1013/1014 is 3.4V (three Ni-Cad batteries). Typical supply current at this voltage is 290 μ A, therefore power dissipation is only one milliwatt per amplifier.

Noise Testing

For applications information on noise testing and calculations, please see the LT1007 or LT1008 data sheet.

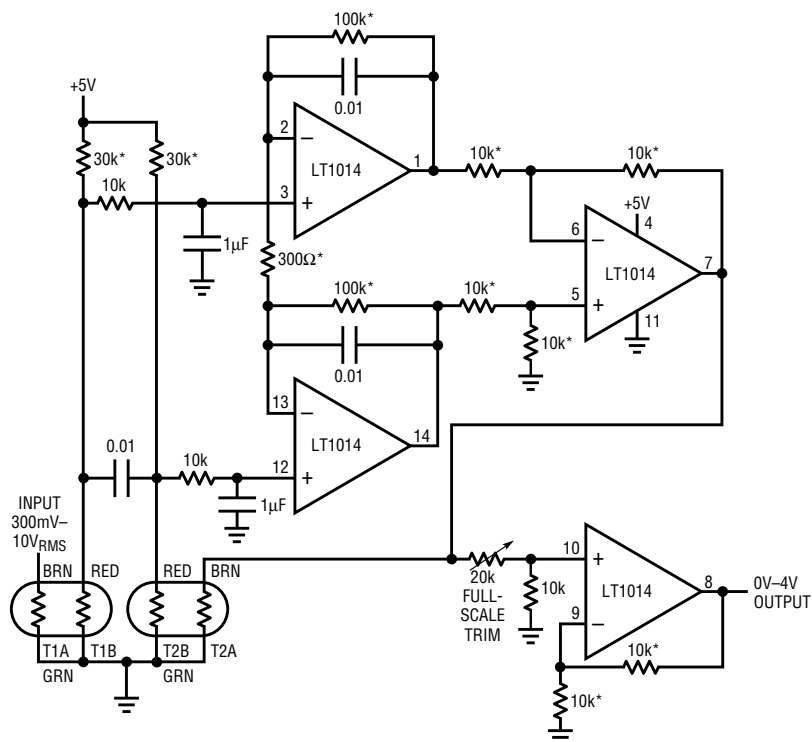
Test Circuit for Offset Voltage and Offset Drift with Temperature



*RESISTOR MUST HAVE LOW THERMOELECTRIC POTENTIAL.
 **THIS CIRCUIT IS ALSO USED AS THE BURN-IN CONFIGURATION, WITH SUPPLY VOLTAGES INCREASED TO ± 20 V.
 $V_O = 1000V_{OS}$

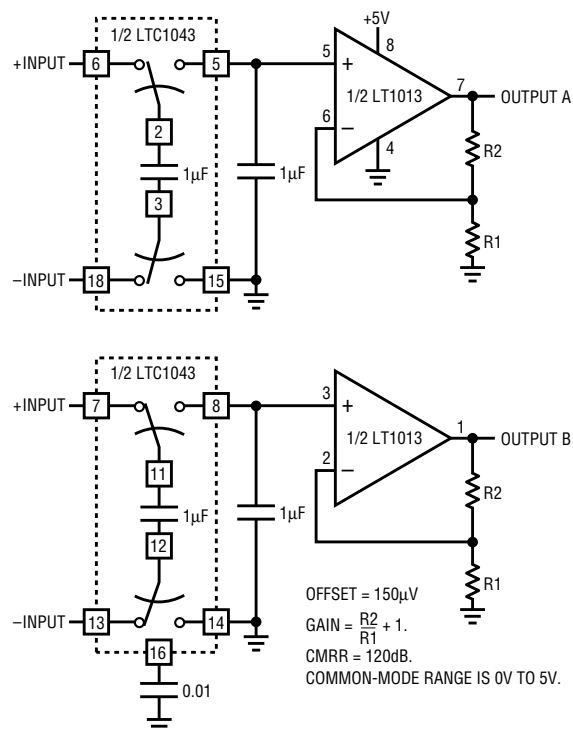
TYPICAL APPLICATIONS

50MHz Thermal rms to DC Converter



2% ACCURACY, DC-50MHz.
 100:1 CREST FACTOR CAPABILITY.
 * 0.1% RESISTOR.
 T1-T2 = YELLOW SPRINGS INST. CO. THERMISTOR COMPOSITE #44018.
 ENCLOSE T1 AND T2 IN STYROFOAM.
 7.5mW DISSIPATION.

5V Single Supply Dual Instrumentation Amplifier



OFFSET = 150 μ V
 GAIN = $\frac{R_2}{R_1} + 1$.
 CMRR = 120dB.
 COMMON-MODE RANGE IS 0V TO 5V.

27Ω
1W

10k*

0.01μF

A1
LT1014

+15V

-15V

2k*

500pF

Q1
2N6533

220

15k

150k*

33k

1k
ZERO FLOW

A2
LT1014

1000pF

2k

3.3k

-15V

Q2-Q5
CA3046
PIN 3 TO -15V

Q2

Q3

Q4

Q5

150k*

12k

1μF

10M
RESPONSE
TIME
ADJUST

2M
FULL-
SCALE
FLOW

500k

1μF

A3
LT1014

9

10

8

A4
LT1014

13

12

14

0V-10V =
0-1000 FEET/MINUTE

REMOVE LAMP'S GLASS ENVELOPE FROM 328 LAMP.
A1 SERVO'S #328 LAMP TO CONSTANT TEMPERATURE.
A2-A3 FURNISH LINEAR OUTPUT vs FLOW RATE.
*1% RESISTOR.



* 1% FILM RESISTOR.
 ** SUPPLIED WITH YSI THERMISTOR NETWORK.
 T1, T2 YSI THERMISTOR NETWORK = #44201.
 FLOW IN PIPE IS INVERSELY PROPORTIONAL TO
 RESISTANCE OF T1-T2 TEMPERATURE DIFFERENCE.
 A1-A2 PROVIDE GAIN. A3-A4 PROVIDE LINEARIZED
 FREQUENCY OUTPUT.

[illegible][illegible]

The circuit diagram illustrates a motor speed control system. It features two LT1013 op-amps, A1 and A2. Op-amp A1 is configured as a duty cycle modulator, with its non-inverting input connected to a +5V supply through a 1kΩ resistor and a 47μF capacitor. Its inverting input is connected to a feedback network consisting of a 100kΩ resistor and a 1MΩ resistor. The output of A1 drives the base of a 2N3904 transistor (Q1). Op-amp A2 is configured as a motor back EMF sampler, with its non-inverting input connected to the motor's back EMF through a 2kΩ resistor. Its inverting input is connected to a feedback network consisting of a 6.8MΩ resistor and a 0.068μF capacitor. The output of A2 drives the base of a 2N5023 transistor (Q2). The motor is connected to a +5V supply through a 1kΩ resistor and a 47μF capacitor. The motor's back EMF is sampled by A2. The motor is a Canon-FN30-R13N1B.

COMPONENTS:

- Op-amps: A1, A2 (1/2 LT1013)
- Transistors: Q1 (2N3904), Q2 (2N5023), Q3 (2N5023)
- Diodes: 1N4001, 1N4148
- Resistors: 330k, 100k, 1M, 2k, 82Ω, 1k, 6.8M, 3.3M, 2k, 0.47, 0.068
- Capacitors: 0.47, 47

NOTES:

- MOTOR = CANON-FN30-R13N1B.
- A1 DUTY CYCLE MODULATES MOTOR.
- A2 SAMPLES MOTORS BACK EMF.

The circuit diagram shows a 1-wire digital-to-analog converter (DALE) using two LT1013 comparators and a transformer. The input is a TTL signal, which is connected to the non-inverting input of the first LT1013 comparator. The inverting input of the first comparator is connected to the secondary of a transformer (DALE #TC-10-04). The transformer's primary is connected to a +5V supply and a 100k resistor. The secondary is connected to a 100k resistor and a 1N4148 diode. The output of the first comparator is connected to the inverting input of the second LT1013 comparator. The non-inverting input of the second comparator is connected to a +5V supply and a 100k resistor. The output of the second comparator is connected to the base of a 2N2222 transistor, which is connected to a 100k resistor and a 6.19k resistor. The emitter of the transistor is connected to ground. The collector of the transistor is connected to a 100k resistor and a 6.19k resistor. The output of the circuit is a 21V signal with a 600μs RC time constant.

MEETS ALL V_{pp} PROGRAMMING SPECS WITH NO TRIMS AND RUNS OFF 5V SUPPLY—NO EXTERNAL HIGH VOLTAGE SUPPLY REQUIRED. SUITABLE FOR BATTERY POWERED USE (600μA QUIESCENT CURRENT). *1% METAL FILM.

*1% METAL FILM RESISTOR
SENSOR = CALECTRO-GC ELECTRONICS #J4-807 OR FIGARO #813

LT1004 1.2V

390k*

2.7k

A3 LT1014

0.033

100k*

10μF

LTC1044

SENSOR

A1 LT1014

5k 1000ppm TRIM

A2 LT1014

1000pF

CA3046

Q1 Q2 Q3 Q4

1N4148 (4)

A4 LT1014

74C04

74C04

74C04

1N4148

OUTPUT
500ppm-10,000ppm
50Hz→1kHz

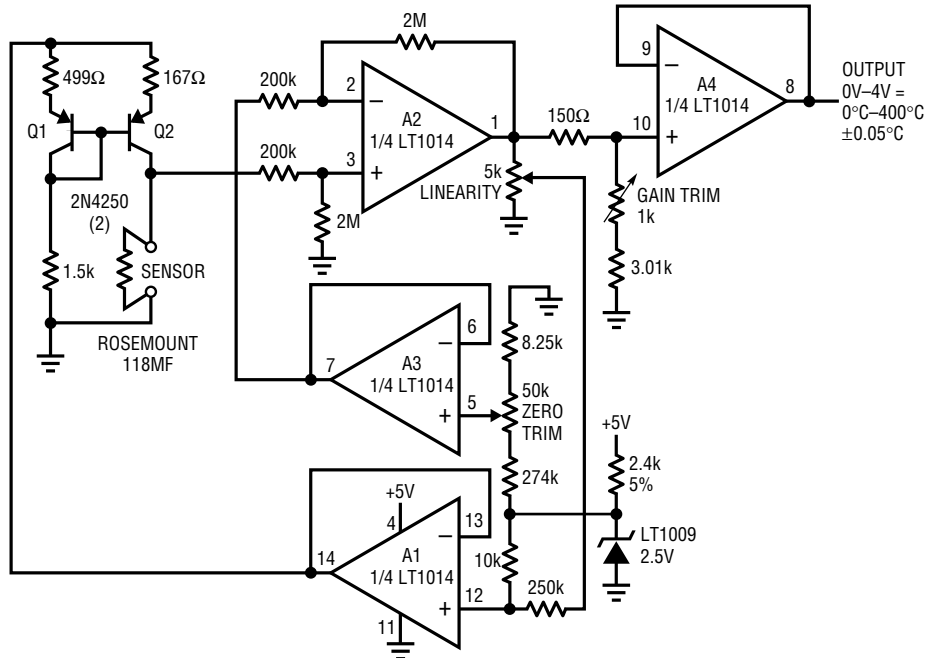
2k

L = DALE TE-3/Q3/TA.
SHORT CIRCUIT CURRENT = 30mA.
≈ 75% EFFICIENCY.
SWITCHING PREREGULATOR CONTROLS DROP ACROSS FET TO 200mV.

† 12-BIT ACCURACY.
 * 1% FILM.
 T1 = PICO-31080.

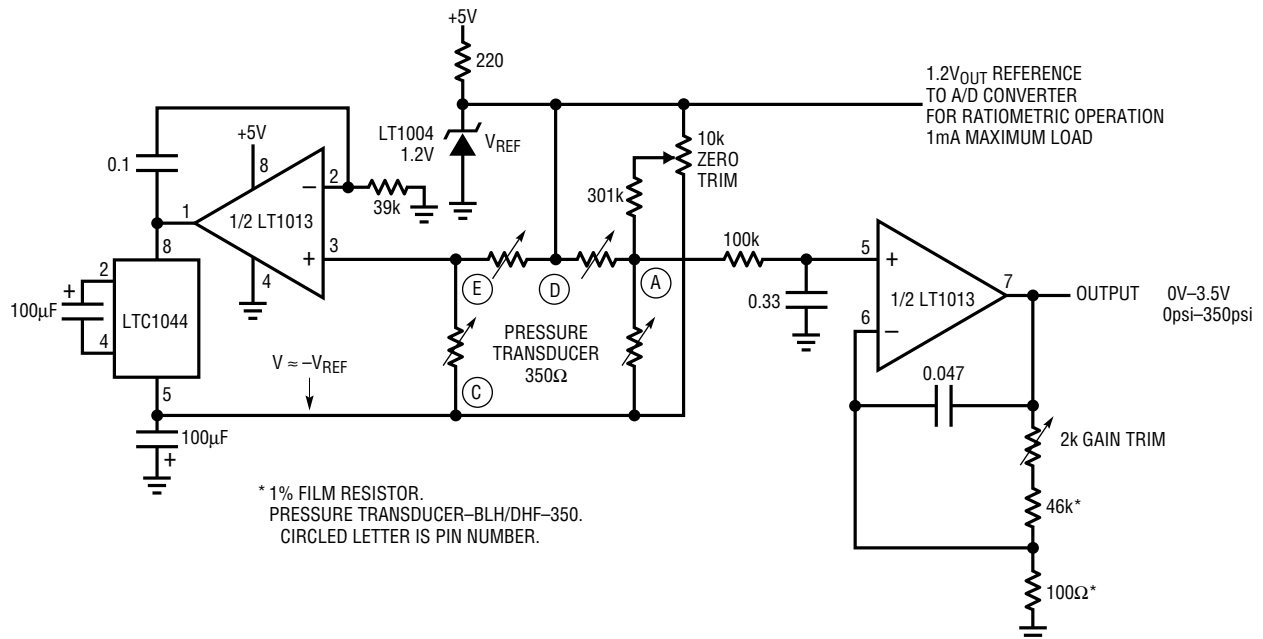
TYPICAL APPLICATIONS

5V Powered, Linearized Platinum RTD Signal Conditioner



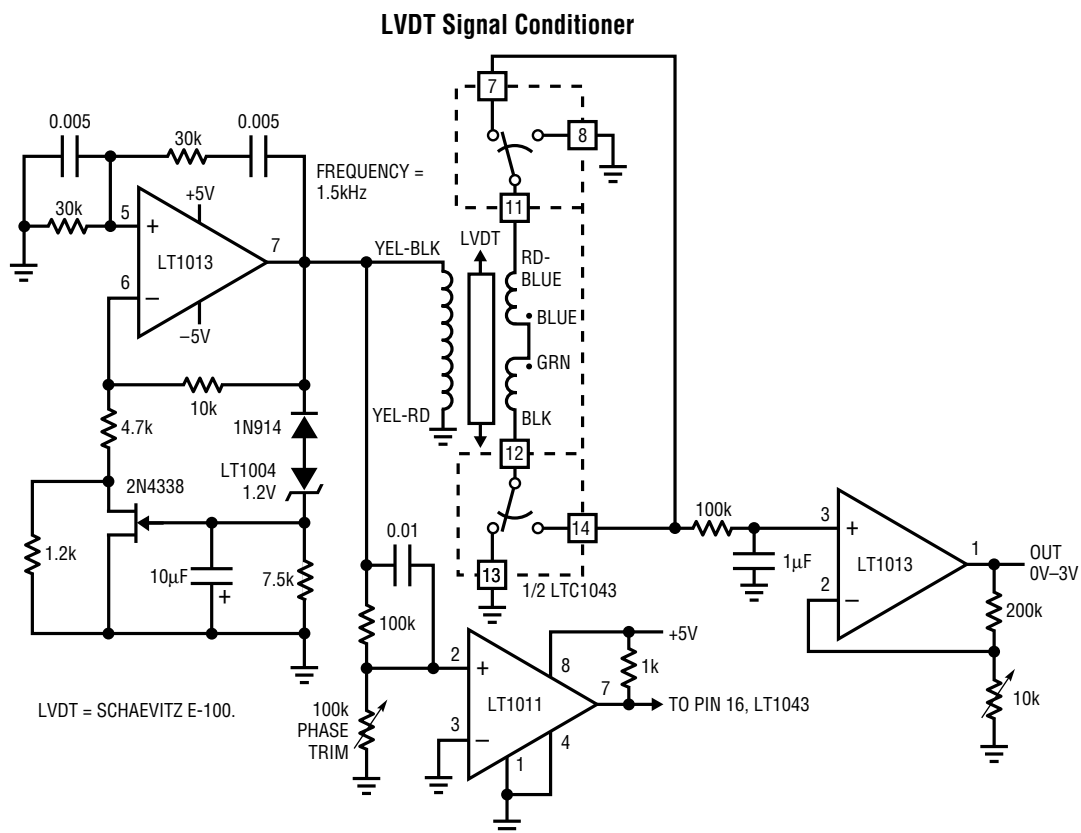
ALL RESISTORS ARE TRW-MAR-6 METAL FILM.
 RATIO MATCH 2M-200K $\pm 0.01\%$.
 TRIM SEQUENCE:
 SET SENSOR TO 0 ° VALUE.
 ADJUST ZERO FOR 0V OUT.
 SET SENSOR TO 100 °C VALUE.
 ADJUST GAIN FOR 1.000V OUT.
 SET SENSOR TO 400 °C.
 ADJUST LINEARITY FOR 4.000V OUT, REPEAT AS REQUIRED.

Strain Gauge Bridge Signal Conditioner

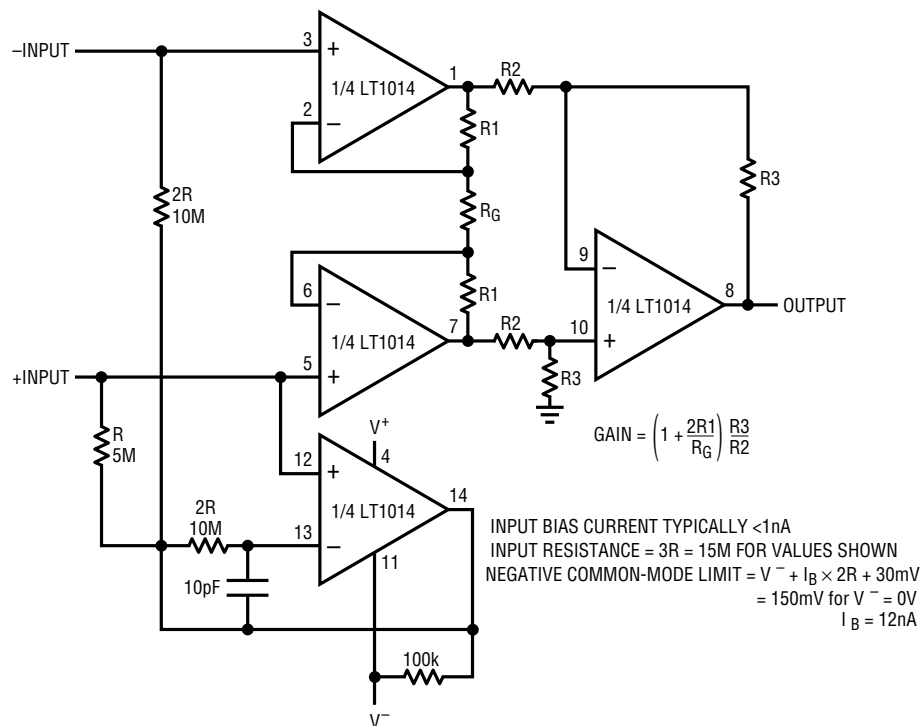


* 1% FILM RESISTOR.
 PRESSURE TRANSDUCER-BLH/DHF-350.
 CIRCLED LETTER IS PIN NUMBER.

TYPICAL APPLICATIONS



Triple Op Amp Instrumentation Amplifier with Bias Current Cancellation



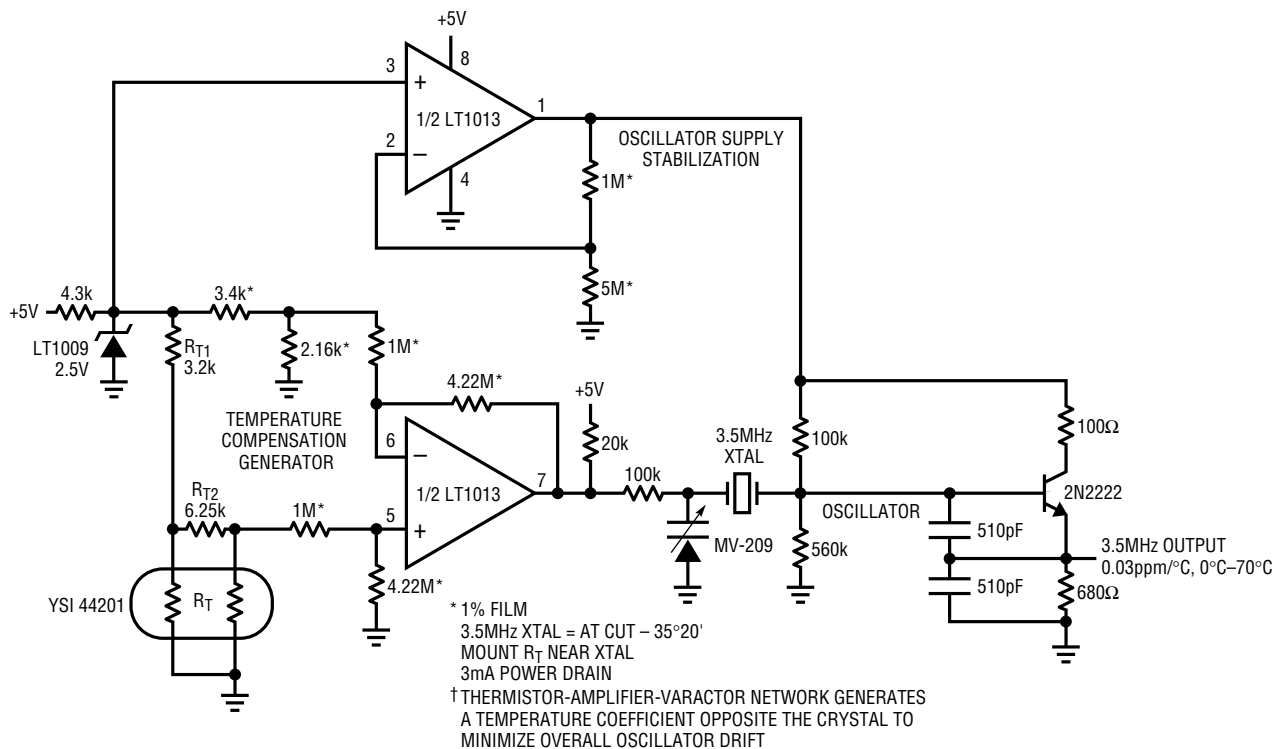
0.009V DROPOUT AT 5mA OUTPUT.
0.108V DROPOUT AT 100mA OUTPUT.
 $I_{\text{QUIESCENT}} = 850\mu\text{A}$.

$I_{OUT} = \frac{V_{IN}}{100\Omega}$
 FOR BIPOLAR OPERATION,
 RUN BOTH ICs FROM
 A BIPOLAR SUPPLY.

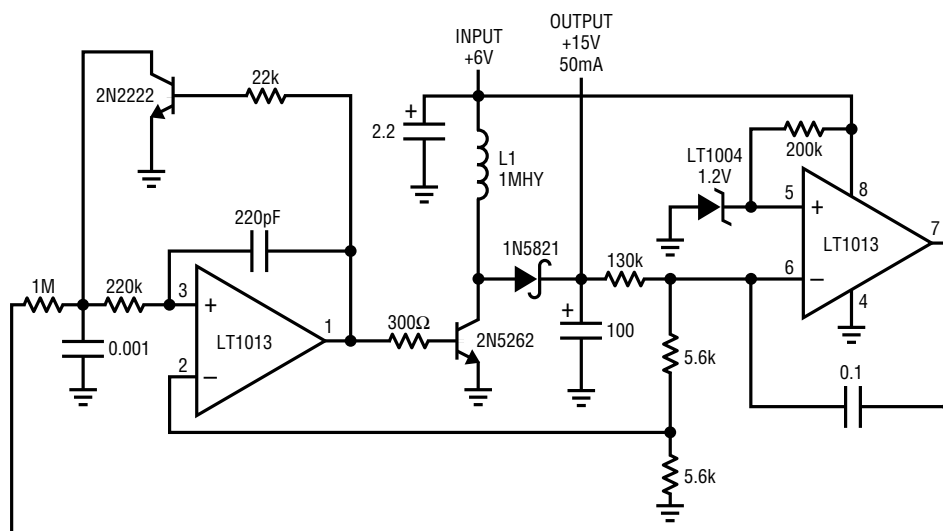
[illegible]

TYPICAL APPLICATIONS

Low Power, 5V Driven, Temperature Compensated Crystal Oscillator (TXCO)[†]

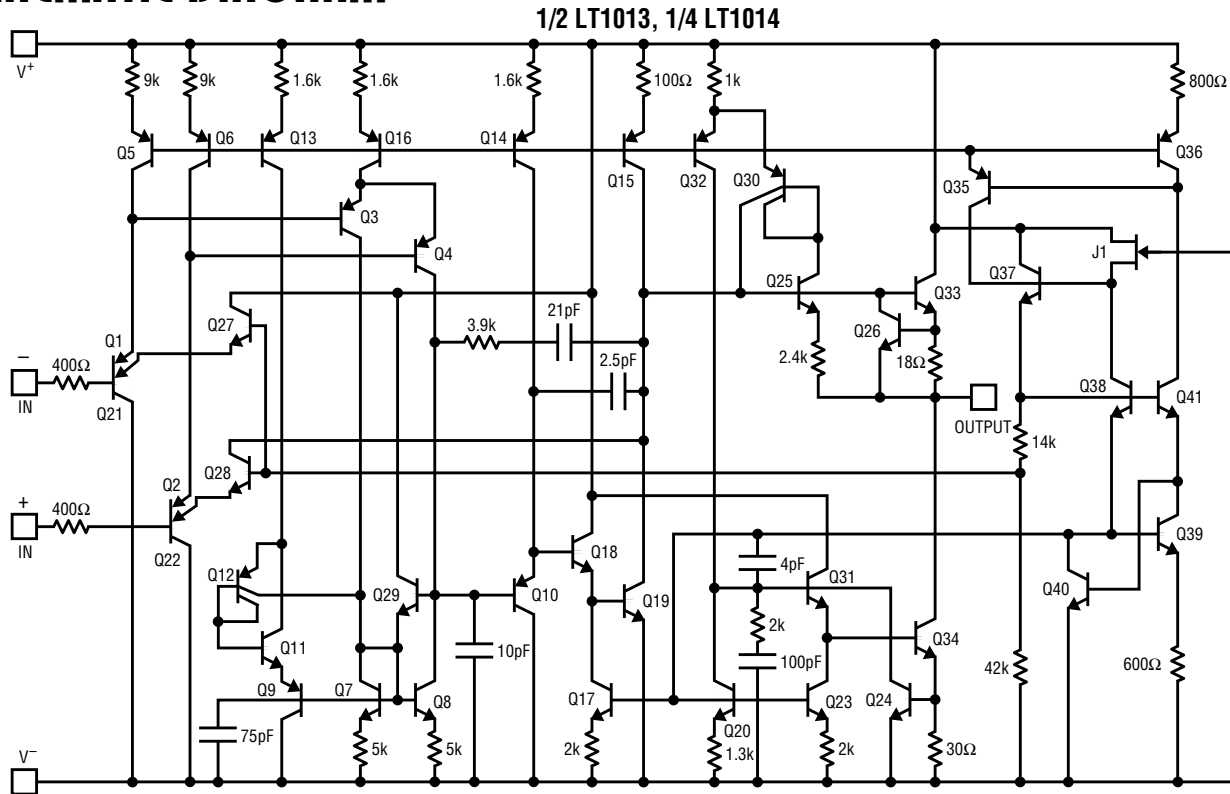


Step-Up Switching Regulator for 6V Battery



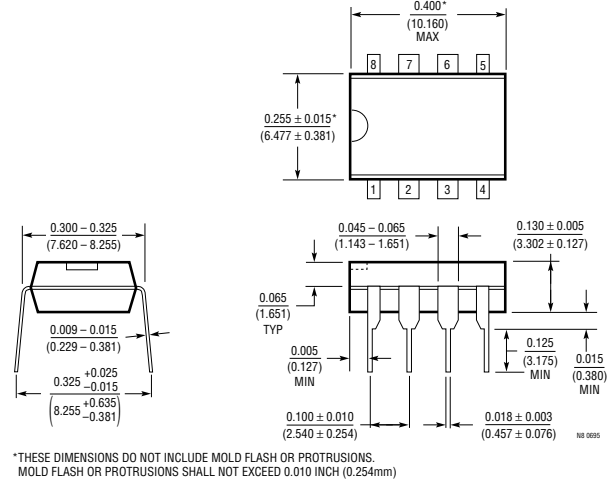
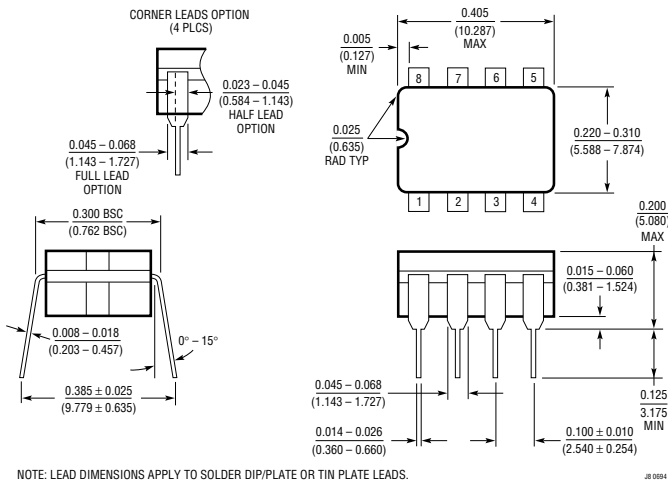
LT = AIE-VERNITRON 24-104
78% EFFICIENCY

SCHEMATIC DIAGRAM



J8 Package
8-Lead CERP (Narrow 0.300, Hermetic)
 (LTC DWG # 05-08-1110)

N8 Package
8-Lead PDIP (Narrow 0.300)
 (LTC DWG # 05-08-1510)

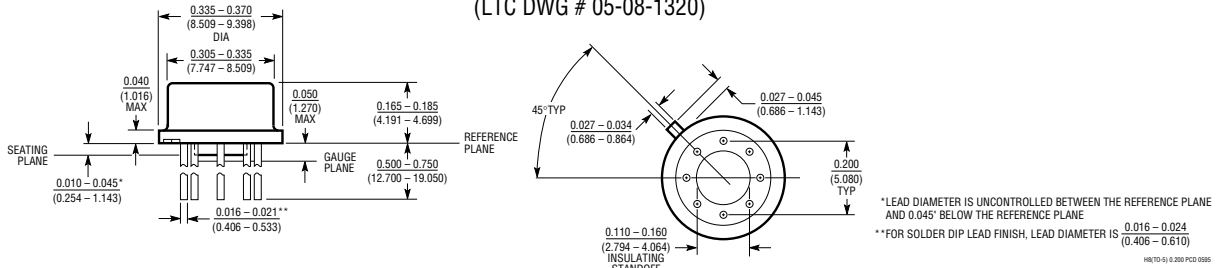


T_{jmax}	θ_{ja}
150°C	100°C/W

T_{jmax}	θ_{ja}
100°C	130°C/W

PACKAGE DESCRIPTION

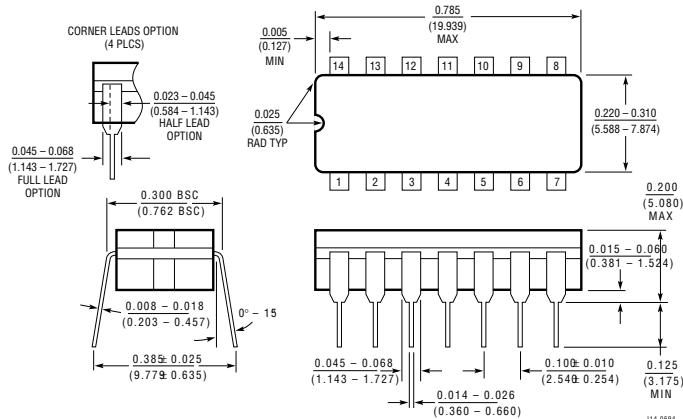
H Package 8-Lead TO-5 Metal Can (0.200 PCD) (LTC DWG # 05-08-1320)



NOTE: DIMENSIONS IN INCHES (MILLIMETERS)

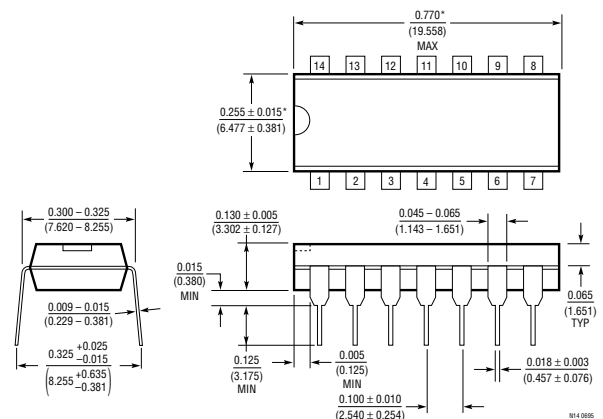
T_{jmax} 150°C	θ_{ja} 150°C/W	θ_{jc} 45°C/W
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J Package 14-Lead CERPDP (Narrow 0.300, Hermetic) (LTC DWG # 05-08-1110)



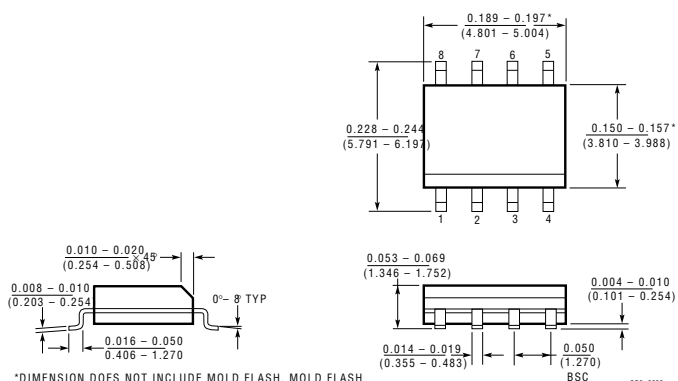
T_{jmax} 150°C	θ_{ja} 100°C/W
---------------------	--------------------------

N Package 14-Lead PDIP (Narrow 0.300) (LTC DWG # 05-08-1510)



T_{jmax} 100°C	θ_{ja} 100°C/W
---------------------	--------------------------

S8 Package 8-Lead Plastic Small Outline (Narrow 0.150) (LTC DWG # 05-08-1610)



SW Package 16-Lead Plastic Small Outline (Wide 0.300) (LTC DWG # 05-08-1620)

